

Advanced Signal Processing 1

VDSL

Very-high-bit-rate Digital Subscriber Lines

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Outline

- VDSL Applications, Goals
- Network Architecture
- CAP
- Achievable Bit-Rates and Limitations
 - Noise
 - Compatibility
- CAP vs. DMT
- Standard development
- Conclusion

VDSL Applications



Application	Downstream	
distance learning	384 kbps–1.5 Mbps	
telecommuting	1.5 Mbps–3.0 Mbps	
multiple digital TV	6.0 Mbps–24.0 Mbps	64 kbps–640 kbps
Internet access	400 kbps–1.5 Mbps	128 kbps–640 kbps
Web hosting	400 kbps–1.5 Mbps	400 kbps–1.5 Mbps
video conferencing	384 kbps–1.5 Mbps	384 kbps–1.5 Mbps
video on demand	6.0 Mbps–18.0 Mbps	64 kbps–128 kbps
interactive video	1.5 Mbps–6.0 Mbps	128 kbps–1.5 Mbps
telemedicine	6.0 Mbps	384 kbps–1.5 Mbps
high-definition TV	16 Mbps	64 kbps

Goals

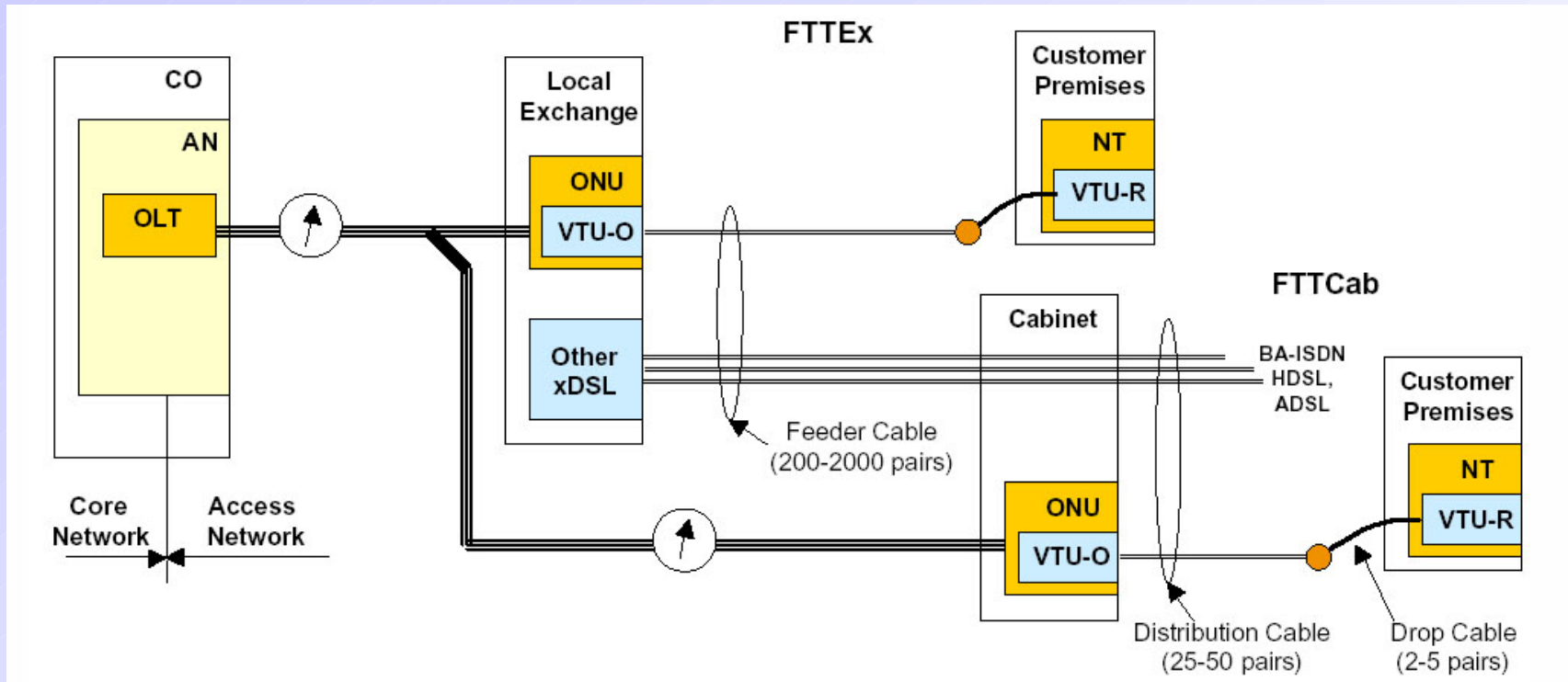
■ Performance Requirements

- Noise Margin $> +6$ dB
- Bit Error ratio of < 1 in 10^7
- Data rate (Mbps)
 - asymmetric 22/3 (NA) 23/4 (Europe)
 - symmetric 13/13 (NA) 28/28 (Europe)
- Latency fast/slow path 1ms/20ms
- POTS or BA-ISDN life-line over the same pair
- Power Consumption < 1 W at the Cabinet

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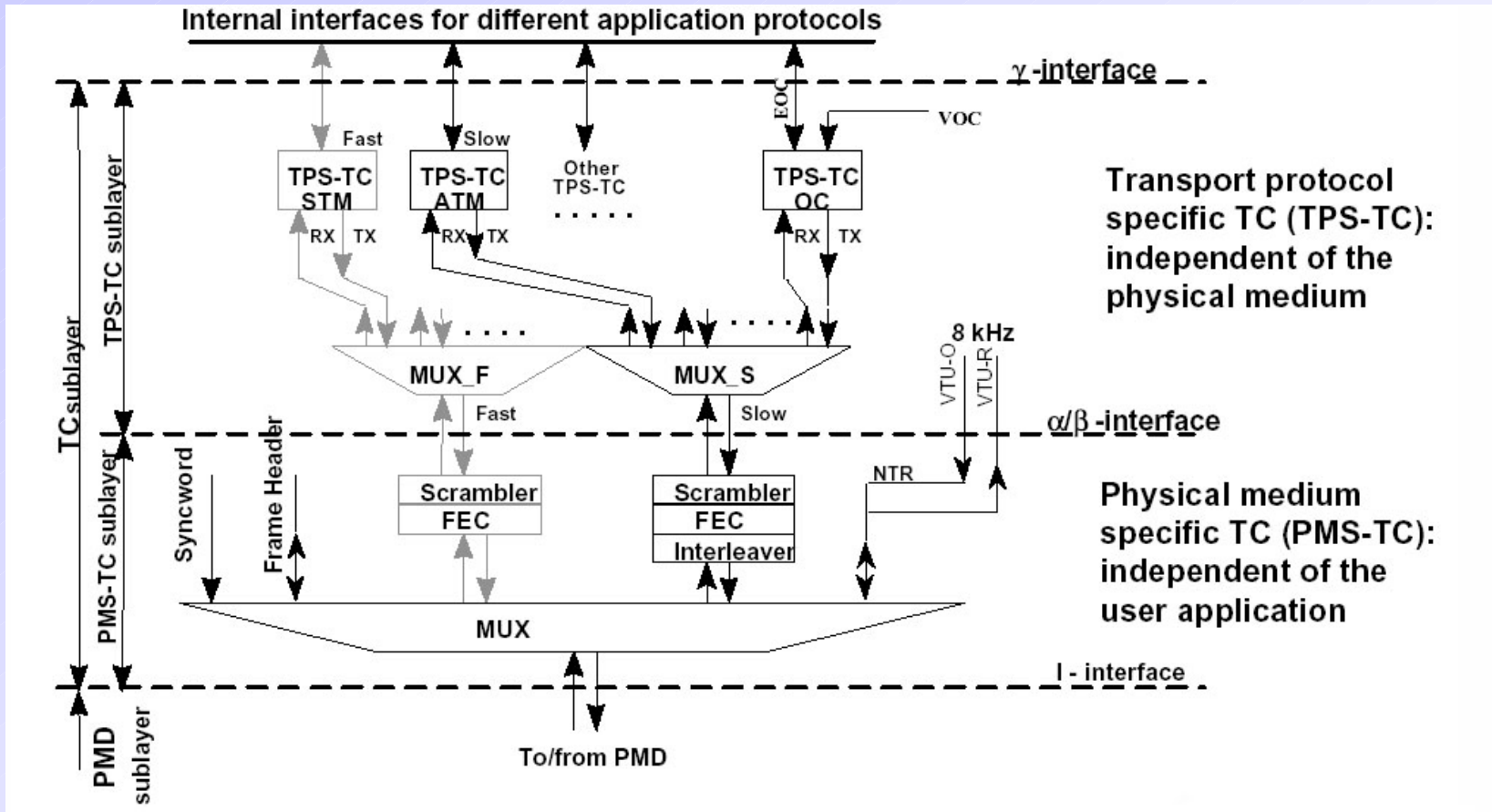
General network architecture



Abbreviations:

AN	Access Network	FTTEx	Fiber-To-The Exchange
ONU	Optical Network Unit	CO	Central Office
VTU	VDSL Transmitting Unit	OLT	Optical Line Termination

TC sublayer architecture

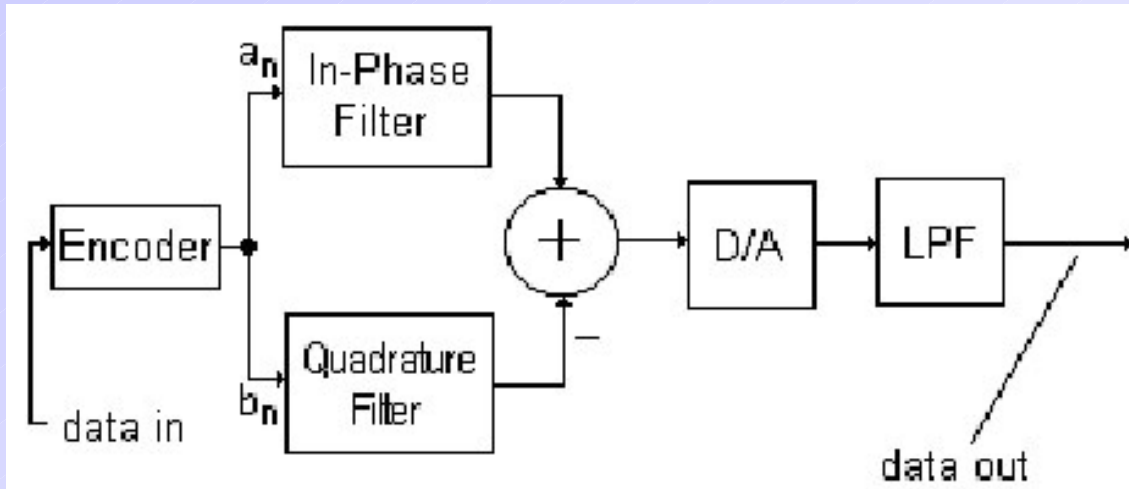


TC Transmission Convergence

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CAP Transmitter



256-CAP

$$A_n = a_n + jb_n$$

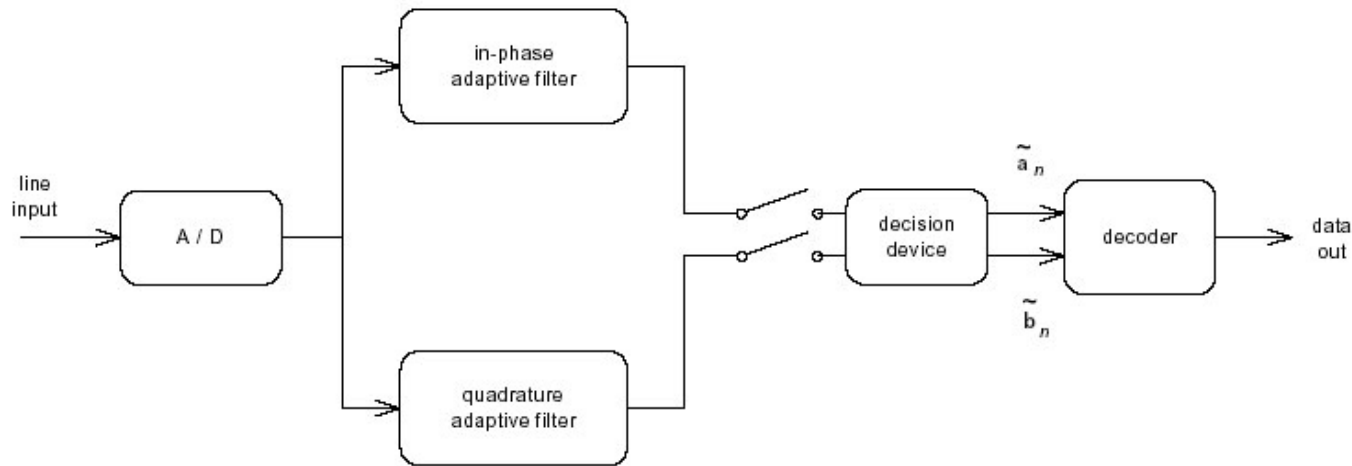
Impulse responses $h(t)$ and $h'(t)$ form a Hilbert pair

$$h(t) = g(t) \cos(2\pi f_c t) \quad h'(t) = g(t) \sin(2\pi f_c t)$$

Transmitted Signal

$$s(t) = \sum_{n=-\infty}^{\infty} [a_n h(t - nT) - b_n h'(t - nT)]$$

CAP Receiver



Conceptual CAP Receiver

- T/I fractionally spaced linear equalizer (FSLE)
- outputs FSLE \rightarrow symbol rate $1/T \rightarrow$ decision device

Recover symbols

■ Linear filtering step:

$$h(t) \otimes f(t) = p(t) \rightarrow h'(t) \otimes f(t) = \tilde{p}(t)$$

■ Impulse Responses of adaptive filters:

$$e_{II} = -\tilde{e}_I$$

■ Output of adaptive filters:

$$s_I = \sum_{n=-\infty}^{\infty} [(a_n p(t - nT) - b_n \tilde{p}(t - nT))]$$

$$s_{II} = \sum_{n=-\infty}^{\infty} [(b_n p(t - nT) - a_n \tilde{p}(t - nT))]$$

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How can we achieve such Very-High-Bit-Rates?

■ Channel Capacity:

$$C = \int_{f_1}^{f_2} \log_2 \left(1 + \frac{S(f)}{N(f)} \right)$$



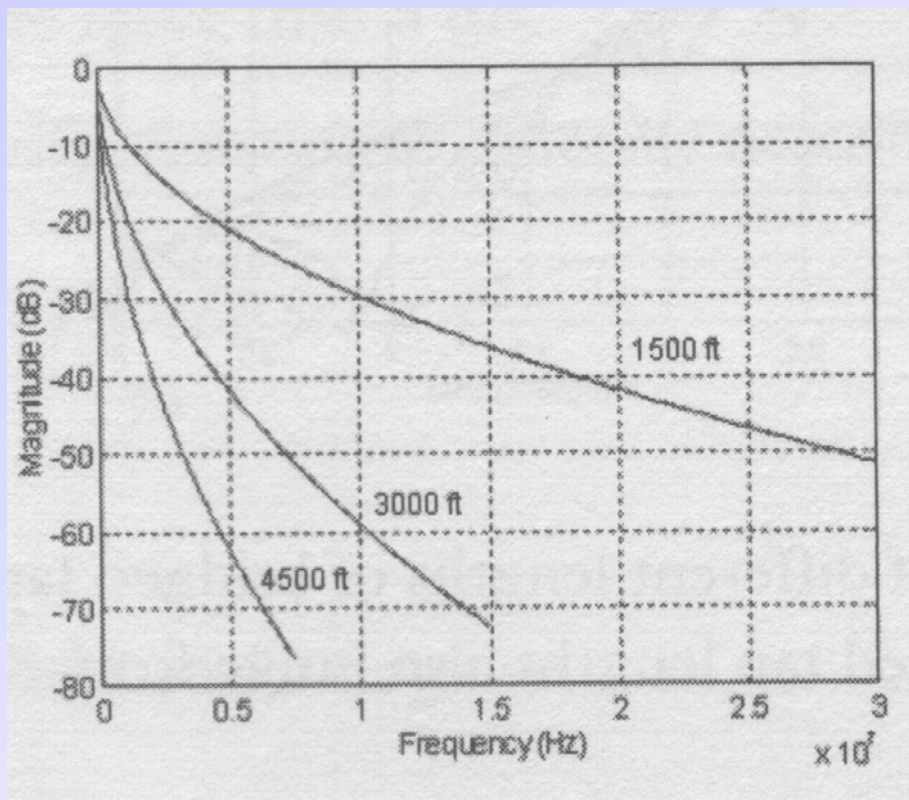
- Expanding signal bandwidth
- High signal-to-noise ratio
- Sophisticated Coding and Modulation

Limiting factors

- Insertion loss
 - Especially for higher frequencies
- Noise
 - Background noise
 - Crosstalk noise
 - Impulse noise
- Compatibility
 - Existing services (e. g. POTS, ISDN, ADSL)
- Radio Frequency Interference
 - Standard amateur and broadcast radio bands
- Economic limitations
 - Power consumption at the Cabinet
 - Cheap and robust solution

Insertion Loss

Insertion Loss of a 24-gauge twisted pair loop



Transfer Function:

$$H(d, f) = e^{-d\gamma(f)} = e^{-d\alpha(f)} e^{-jd\beta(f)}$$

Propagation Loss in dB/m:

$$L_p(f) = -20 \log |H(1, f)|$$

$$L_p(f) \approx 8.686(a\sqrt{f} + bf)$$

Noise Model

Background Noise

- White noise -140 dBm/Hz

Crosstalk Noise

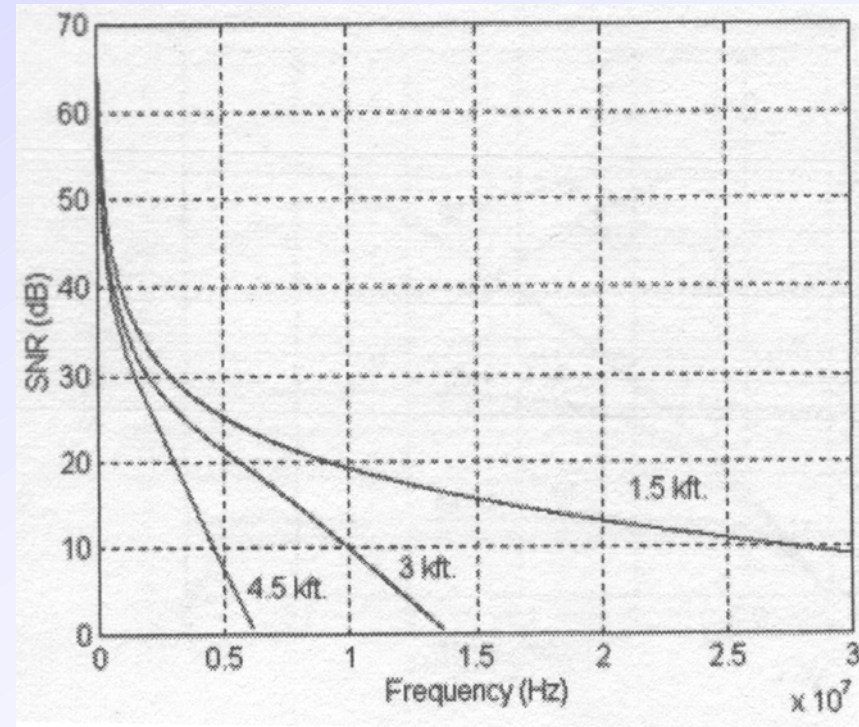
$$\frac{S(f)}{N_F(f)} = \frac{Q(f)|H(f)|^2}{Q(f)FEXT(f)} = \frac{1}{klf^2}$$

$S(f)$ Received signal power density

$Q(f)$ Transmit power spectral density

$H(f)$ Channel transfer function

$FEXT(f)$ power sum transfer function



SNR for 24-gauge TP-Loop

Channel Capacity

■ Channel Capacity Results:

- Length of TP loop in m \rightarrow Mbps
- 500/1000/1500 \rightarrow 160/68/25

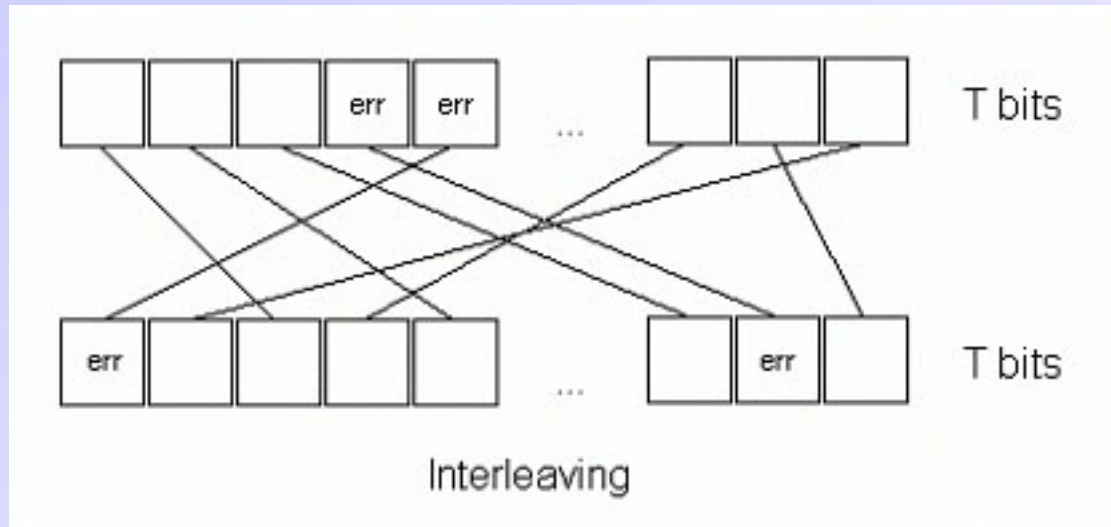
■ 6 dB performance margin

- 500/1000/1500 \rightarrow 108/44/18

■ Realized transmission throughputs

- Lower due to allocation of guardbands and not using the whole spectrum
- Europe 28Mbps/28Mbps

Impulse Noise



■ Interleaving

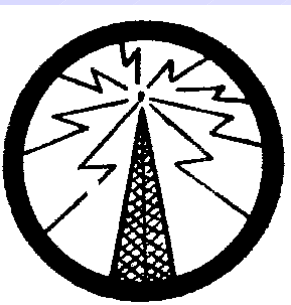
- Temporal permutation of Bits
- Errors spread over block
- Only for slow path

Radio Frequency Interference



■ RF Egress

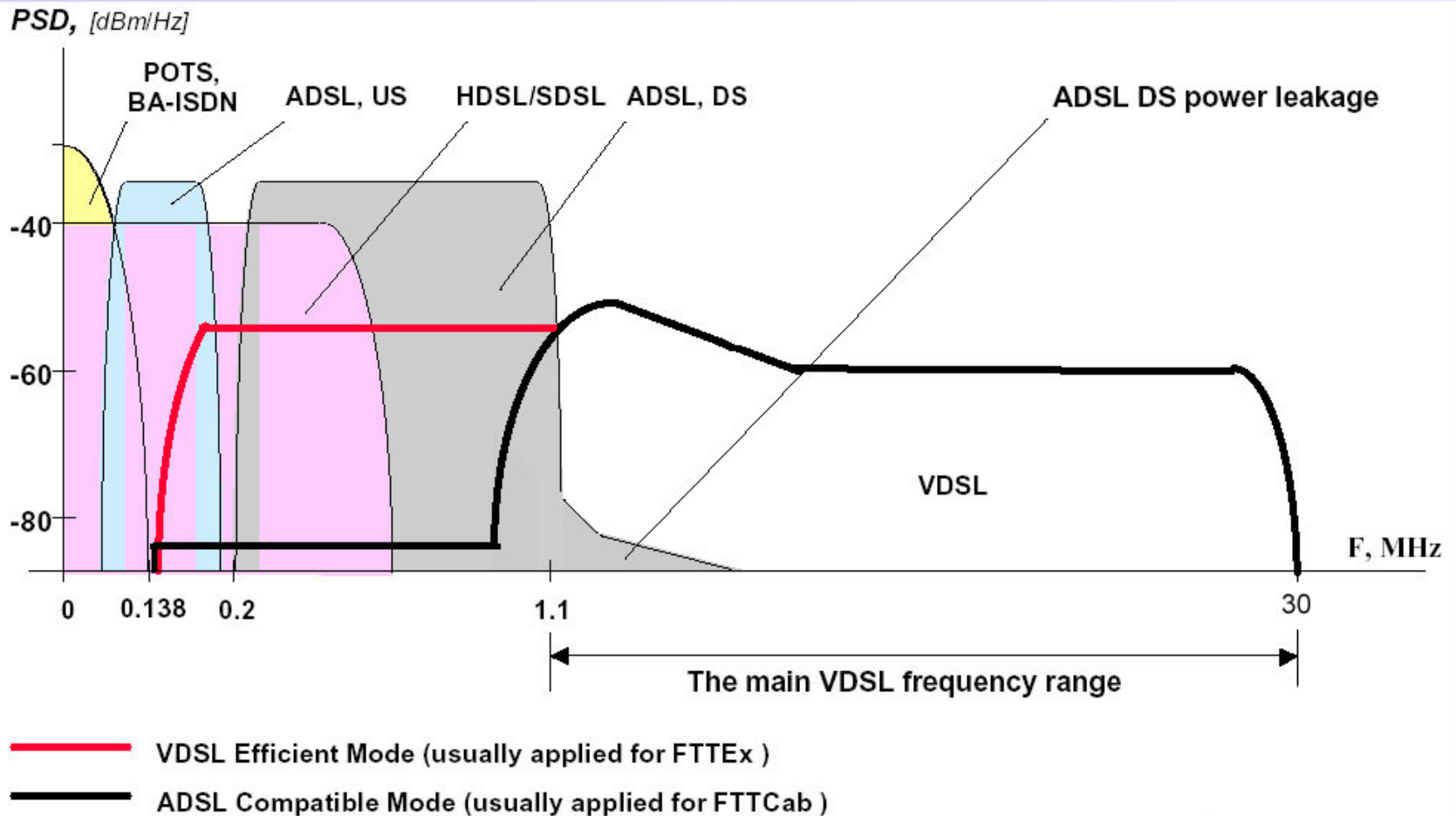
- Caused by drop cable connection
- At customer premises
- HAM radio frequencies
 - CAP → Set notch filters
 - DMT → switch off subcarrier



■ RF Ingress

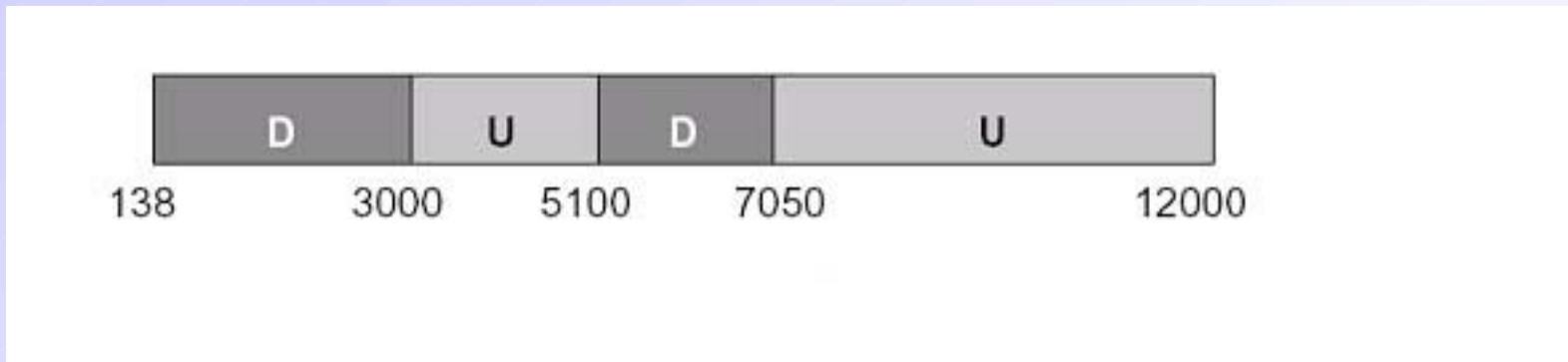
- CAP → equalizer
- DMT → switch off subcarrier

Spectral Compatibility



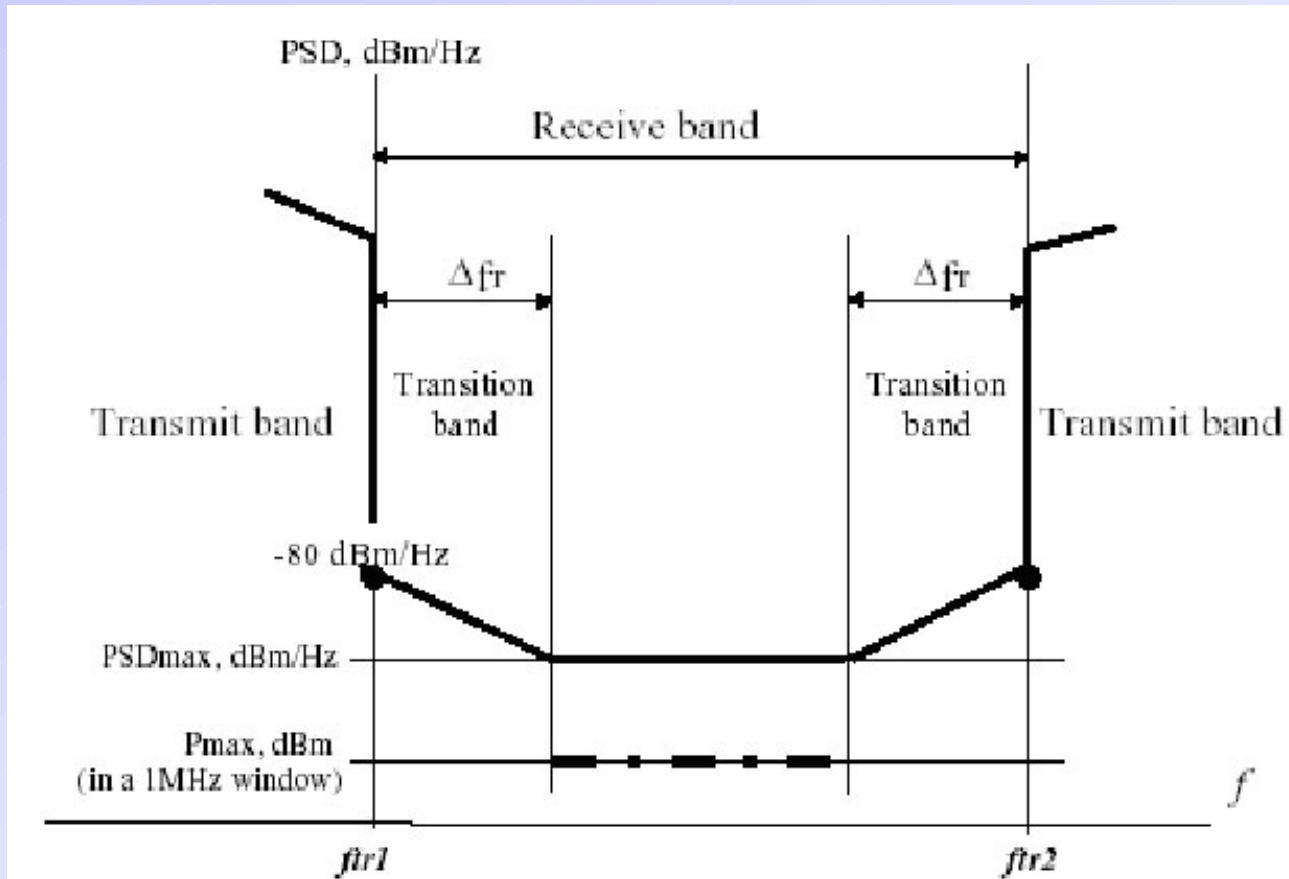
Transmit spectra

■ Band allocation Europe



- D ... Downstream transmission
- U ... Upstream transmission

Out-of-Band PSD mask



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CAP vs DMT I

- DMT – VDSL Alliance
 - Texas Instruments and Alcatel
- CAP - Coalition
 - Lucent, Broadcom, Infineon

CAP vs DMT II

■ DMT

- 😊
 - No error propagation
 - Subchannels degrade → redistributing data to other subchannels
- ☹️
 - Relies on signal analysis techniques → reliability in the field?
 - Requires 0.18 micron processes to meet the power constraints

■ CAP

- 😊
 - 0.25 micron processes to meet the power constraints
 - Relies on well-known and extensively field-proven digital signal processing algorithms
- ☹️
 - Equalizer can create errors that propagate
 - Equalizers with short, low-complexity filters. do not perform well on difficult channels

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VDSL Standard

- Europe (ETSI TM6)
 - 2 parts: Functional requirements and Transceiver specification
 - SCM, MCM technologies are specified as possible implementations
- North America (ANSI T1E1.4)
 - 3 parts: Functional requirements, SCM, MCM Transceiver specification
- International (ITU-T)
 - only Functional requirements

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Conclusion

- VDSL is
- ... a well developed technology
- ... at the last stages of standardization
- ... a multiservice architecture
- ... designed to operate in the presence of all kinds of impairments in copper pairs
- ... spectrally compatible with other xDSL
- And will have a great market potential